

BGT24ATR11

Silicon Germanium 24 GHz Transceiver MMIC

Data Sheet

Revision 3.0, 2013-09-10

Edition 2013-09-10

**Published by
Infineon Technologies AG
81726 Munich, Germany**

**© 2013 Infineon Technologies AG
All Rights Reserved.**

Legal Disclaimer

The information given in this document shall in no event be regarded as a guarantee of conditions or characteristics. With respect to any examples or hints given herein, any typical values stated herein and/or any information regarding the application of the device, Infineon Technologies hereby disclaims any and all warranties and liabilities of any kind, including without limitation, warranties of non-infringement of intellectual property rights of any third party.

Information

For further information on technology, delivery terms and conditions and prices, please contact the nearest Infineon Technologies Office (www.infineon.com).

Warnings

Due to technical requirements, components may contain dangerous substances. For information on the types in question, please contact the nearest Infineon Technologies Office.

Infineon Technologies components may be used in life-support devices or systems only with the express written approval of Infineon Technologies, if a failure of such components can reasonably be expected to cause the failure of that life-support device or system or to affect the safety or effectiveness of that device or system. Life support devices or systems are intended to be implanted in the human body or to support and/or maintain and sustain and/or protect human life. If they fail, it is reasonable to assume that the health of the user or other persons may be endangered.

BGT24ATR11 Silicon Germanium 24 GHz Transceiver MMIC

Revision History: 2013-09-10, Revision 3.0

Previous Revision: 2012-02-08, Revision 2.4

| Page | Subjects (major changes since last revision) |
|------|--|
| all | Update parameter |
| | |
| | |
| | |
| | |

Trademarks of Infineon Technologies AG

AURIX™, BlueMoon™, C166™, CanPAK™, CIPOS™, CIPURSE™, COMNEON™, EconoPACK™, CoolMOS™, CoolSET™, CORECONTROL™, CROSSAVE™, DAVE™, EasyPIM™, EconoBRIDGE™, EconoDUAL™, EconoPIM™, EiceDRIVER™, eupec™, FCOS™, HITFET™, HybridPACK™, I²PRF™, ISOFACE™, IsoPACK™, MIPAQ™, ModSTACK™, my-d™, NovalithIC™, OmniTune™, OptiMOS™, ORIGA™, PRIMARION™, PrimePACK™, PrimeSTACK™, PRO-SIL™, PROFET™, RASICT™, ReverSave™, SatRIC™, SIEGET™, SINDRION™, SIPMOS™, SMARTi™, SmartLEWIS™, SOLID FLASH™, TEMPFET™, thinQ!™, TRENCHSTOP™, TriCore™, X-GOLD™, X-PMU™, XMM™, XPOSSYS™.

Other Trademarks

Advance Design System™ (ADS) of Agilent Technologies, AMBA™, ARM™, MULTI-ICET™, KEIL™, PRIMECELL™, REALVIEW™, THUMB™, μVision™ of ARM Limited, UK. AUTOSAR™ is licensed by AUTOSAR development partnership. Bluetooth™ of Bluetooth SIG Inc. CAT-iq™ of DECT Forum. COLOSSUS™, FirstGPS™ of Trimble Navigation Ltd. EMV™ of EMVCo, LLC (Visa Holdings Inc.). EPCOS™ of Epcos AG. FLEXGO™ of Microsoft Corporation. FlexRay™ is licensed by FlexRay Consortium. HYPERTERMINAL™ of Hilgraeve Incorporated. IEC™ of Commission Electrotechnique Internationale. IrDA™ of Infrared Data Association Corporation. ISO™ of INTERNATIONAL ORGANIZATION FOR STANDARDIZATION. MATLAB™ of MathWorks, Inc. MAXIM™ of Maxim Integrated Products, Inc. MICROTEC™, NUCLEUS™ of Mentor Graphics Corporation. Mifare™ of NXP. MIPI™ of MIPI Alliance, Inc. MIPS™ of MIPS Technologies, Inc., USA. muRata™ of MURATA MANUFACTURING CO., MICROWAVE OFFICE™ (MWO) of Applied Wave Research Inc., OmniVision™ of OmniVision Technologies, Inc. Openwave™ Openwave Systems Inc. RED HAT™ Red Hat, Inc. RFMD™ RF Micro Devices, Inc. SIRIUS™ of Sirius Satellite Radio Inc. SOLARIS™ of Sun Microsystems, Inc. SPANSION™ of Spansion LLC Ltd. Symbian™ of Symbian Software Limited. TAIYO YUDEN™ of Taiyo Yuden Co. TEAKLITE™ of CEVA, Inc. TEKTRONIX™ of Tektronix Inc. TOKO™ of TOKO KABUSHIKI KAISHA TA. UNIX™ of X/Open Company Limited. VERILOG™, PALLADIUM™ of Cadence Design Systems, Inc. VLYNQ™ of Texas Instruments Incorporated. VXWORKS™, WIND RIVER™ of WIND RIVER SYSTEMS, INC. ZETEX™ of Diodes Zetex Limited.

Last Trademarks Update 2010-10-26

Table of Contents

| | | |
|----------|---|----|
| | Table of Contents | 4 |
| | List of Figures | 5 |
| | List of Tables | 6 |
| 1 | Features | 7 |
| 2 | Electrical Characteristics | 9 |
| 2.1 | Absolute Maximum Ratings | 9 |
| 2.2 | Thermal Resistance | 10 |
| 2.3 | ESD Integrity | 10 |
| 2.4 | Measured RF Characteristics | 11 |
| 2.4.1 | Power Supply | 11 |
| 2.4.2 | TX Section | 11 |
| 2.4.3 | RX Section | 14 |
| 2.5 | Temperature Sensor | 15 |
| 2.6 | Power Detector | 15 |
| 3 | Application Circuit and Block Diagram | 16 |
| 3.1 | Application Circuit Schematic | 16 |
| 3.2 | Pin Description | 17 |
| 3.3 | SPI | 18 |
| 3.4 | Application Board and Reflow Profile | 21 |
| 3.5 | Equivalent Circuit Diagram of MMIC Interfaces | 23 |
| 4 | Physical Characteristics | 24 |
| 4.1 | Package Footprint | 24 |
| 4.2 | Reflow Profile | 25 |
| 4.3 | Package Dimensions | 26 |

List of Figures

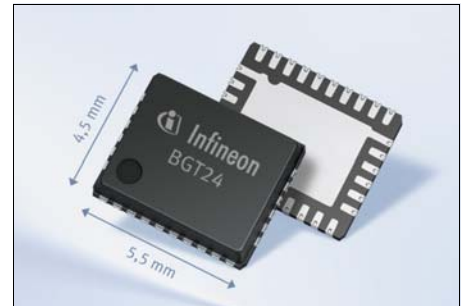
| | | |
|-----------|---|----|
| Figure 1 | BGT24ATR11 Block Diagram | 8 |
| Figure 2 | VCO Tuning Window | 12 |
| Figure 3 | Application Circuit with Chip Outline (Top View) | 16 |
| Figure 4 | Timing Diagram of the SPI | 19 |
| Figure 5 | Cross-Section View of Application Board | 21 |
| Figure 6 | Detail of Compensation Structure (valid for appl. board mat. Ro4350B, 0.254mm acc. to Fig. 5) . | 21 |
| Figure 7 | Application Board Layout | 22 |
| Figure 8 | Equivalent Circuit Diagram of MMIC Interfaces | 23 |
| Figure 9 | Recommended Footprint and Stencil Layout for the VQFN32-9 Package | 24 |
| Figure 10 | Reflow Profile for BGT24ATR11 (VQFN32-9) | 25 |
| Figure 11 | Package Outline (Top, Side and Bottom View) of VQFN32-9 | 26 |
| Figure 12 | Marking Layout VQFN32-9 | 26 |
| Figure 13 | Tape of VQFN32-9 | 27 |

List of Tables

| | | |
|----------|--|----|
| Table 1 | Absolute Maximum Ratings | 9 |
| Table 2 | Thermal Resistance | 10 |
| Table 3 | ESD Integrity | 10 |
| Table 4 | Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, SPI-Bit 4 = high | 11 |
| Table 5 | Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high | 11 |
| Table 6 | Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high | 12 |
| Table 7 | Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high | 14 |
| Table 8 | Typical Characteristics Temperature Sensor $T_A = -40 \dots 125 \text{ }^\circ\text{C}$ | 15 |
| Table 9 | Typical Characteristics Power Detector $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $V_{CC} = 3.3 \text{ V}$ | 15 |
| Table 10 | Bill of Materials | 16 |
| Table 11 | Pin Definition and Function | 17 |
| Table 12 | SPI Data Bit Description | 18 |
| Table 13 | SPI Timing and Logic Levels | 19 |
| Table 14 | Truth Table AMUX | 19 |

1 Features

- 24 GHz ISM band transceiver MMIC
- Qualified according AEC-Q100
- Fully integrated low phase noise VCO
- Switchable prescaler with 1.5 GHz and 23 kHz output
- On chip power and temperature sensors
- Gilbert based homodyne quadrature receiver
- Single ended RF and LO terminals
- Low noise figure NF_{SSB} : 12 dB
- High conversion gain: 26 dB
- High 1 dB input compression point: -12 dBm
- Single supply voltage 3.3 V
- Low power consumption 500 mW
- 200 GHz bipolar SiGe:C technology b7hf200
- Fully ESD protected device
- VQFN-32-9 leadless plastic package incl. LTI feature
- Pb-free (RoHS compliant) package



Description

The BGT24ATR11 is a Silicon Germanium MMIC for signal generation and reception, operating in the 24 GHz ISM band from 24.00 to 24.25 GHz. It is based on a 24 GHz fundamental voltage controlled oscillator. Switchable frequency prescalers are included with output frequencies of 1.5 GHz and 23 kHz. The main RF output delivers typ. 11dBm signal power to feed an antenna and an auxiliary LO output is available to provide LO signal to separate receiver components.

A LNA provides low noise figure and a RC polyphase filter (PPF) is used for LO quadrature phase generation of the homodyne quadrature downconversion mixer. Output power sensors as well as a temperature sensor are implemented for monitoring purposes. The device is controlled via SPI and is manufactured in a 0.18 μ m SiGe:C technology offering a cutoff frequency of 200 GHz. The MMIC is packaged in a 32 pin leadless RoHS compliant VQFN package.

| Product Name | Package | Chip | Marking |
|--------------|----------|-------|------------|
| BGT24ATR11 | VQFN32-9 | T1524 | BGT24ATR11 |

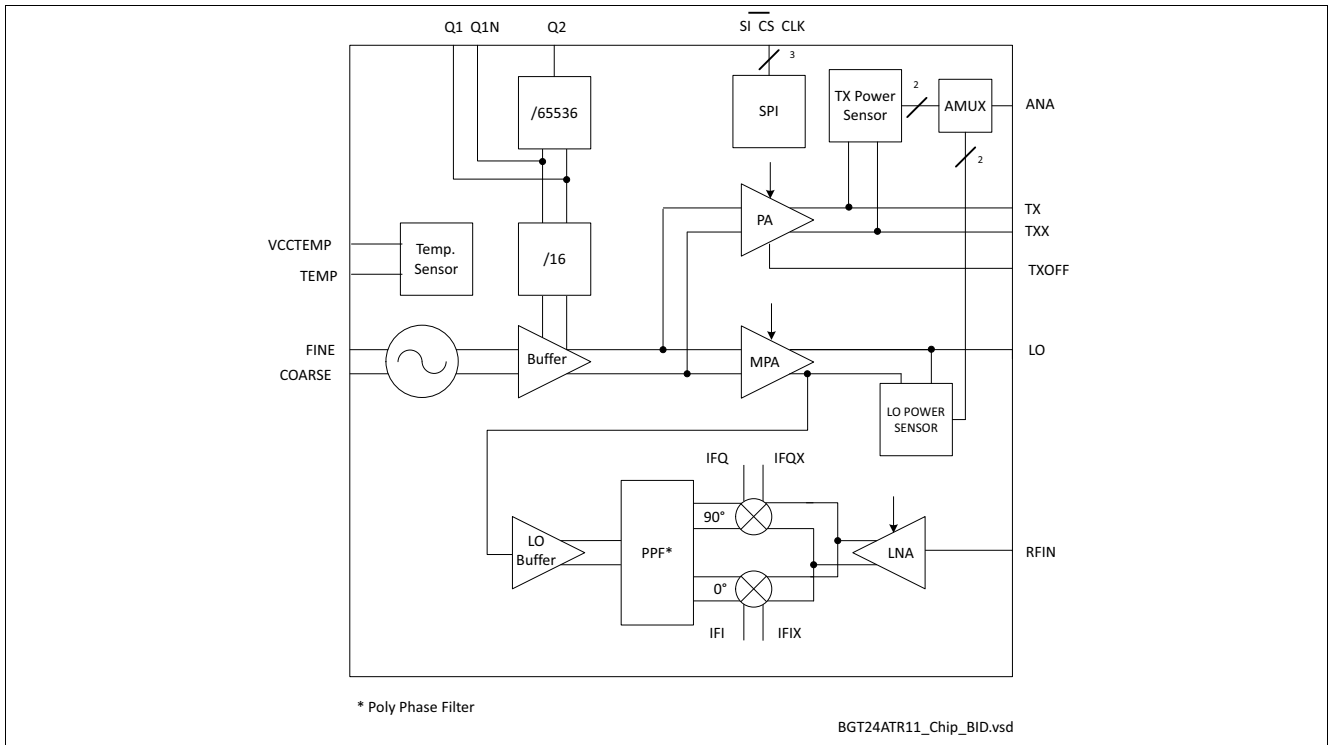


Figure 1 BGT24ATR11 Block Diagram

2 Electrical Characteristics

2.1 Absolute Maximum Ratings

$T_A = -40\text{ °C}$ to 125 °C ; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)¹⁾

Table 1 Absolute Maximum Ratings

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|---|-----------------------|---------|------|------|---------|--|
| | | Min. | Typ. | Max. | | |
| Supply voltage | V_{CC} / V_{CCTEMP} | -0.3 | – | 3.6 | V | – |
| DC voltage at RF Pins TX, TXX, LO, RFIN1, RFIN2 | VDC_{RF} | 0 | – | 0 | V | MMIC provides short circuit to GND for all RF pins |
| DC voltage at Pins IFI, IFIX, IFQ, IFQX | VDC_{IF} | 0 | – | Vcc | V | – |
| DC current into Pins IFI, IFIX, IFQ, IFQX | I_{IF} | -8.5 | – | 3.5 | mA | Max. values indicate current due to short circuit to GND and Vcc resp. |
| DC voltage at Pin ANA | VDC_{ANA} | -0.3 | – | 3.6 | V | – |
| DC current into Pin ANA (Sink) | $I_{ANA\ SINK}$ | 125 | 350 | 500 | μ A | Max. values indicate current due to short circuit to GND and Vcc resp. |
| DC current into Pin ANA (Source) | $I_{ANA\ SOURCE}$ | -7 | – | – | mA | – |
| DC current into Pin VCCTEMP | $I_{VCCTEMP}$ | – | – | 3.5 | mA | Max. values indicate current due to short circuit to GND and Vcc resp. |
| DC voltage at Pin TEMP | VDC_{TEMP} | 0 | – | Vcc | V | – |
| DC current into Pin TEMP | I_{TEMP} | -1 | – | 1.5 | mA | Max. values indicate current due to short circuit to GND and Vcc resp. |
| DC voltage at Pins Q1, Q1N | VDC_{Q1x} | Vcc-0.3 | – | Vcc | V | – |
| DC current into Pins Q1, Q1N | I_{Q1x} | -8 | – | 12 | mA | – |
| DC voltage at Pin Q2 | VDC_{Q2} | -0.3 | – | 3.6 | V | – |
| DC current into Pin Q2 enabled | I_{Q2EN} | -3 | – | 3 | mA | – |
| DC current into Pin Q2 disabled | I_{Q2DIS} | -10 | – | 10 | μ A | – |
| DC voltage at SPI input Pins SI, CLK, \overline{CS} | VDC_{SPIIN} | -0.3 | – | 3.6 | V | – |
| DC current into SPI input Pins SI, CLK, \overline{CS} | I_{SPIIN} | – | – | 3 | mA | – |
| RF input power into Pin RFIN | P_{RF} | – | – | 0 | dBm | – |

1) Not subject to production test, specified by design

Table 1 Absolute Maximum Ratings (cont'd)

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|-----------------------------------|---------------|--------|------|------|--------------|--|
| | | Min. | Typ. | Max. | | |
| DC voltage at Pins Fine, Coarse | V_F, V_C | 0 | – | 5 | V | – |
| DC current into Pins Fine, Coarse | I_F, I_C | -110 | – | 110 | μ A | Positive currents if $V_{TUNE} > V_{CC}$; Negative currents if $0V \leq V_F, V_C \leq V_{CC}$ |
| DC voltage at Pin TXOFF | VDC_{TXOFF} | -0.3 | – | 3.6 | V | – |
| Total power dissipation | P_{DISS} | – | – | 750 | mW | With BIST deactivated |
| Junction temperature | T_J | -40 | – | 155 | $^{\circ}$ C | – |
| Ambient temperature range | T_A | -40 | – | 125 | $^{\circ}$ C | T_A = temperature at package soldering point |
| Storage temperature range | T_{STG} | -40 | – | 150 | $^{\circ}$ C | – |

Attention: Stresses exceeding the max. values listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Maximum ratings are absolute ratings; exceeding only one of these values may cause irreversible damage to the integrated circuit.

2.2 Thermal Resistance

Table 2 Thermal Resistance

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|--|------------|--------|------|------|------|-----------------------|
| | | Min. | Typ. | Max. | | |
| Junction - soldering point ¹⁾ | R_{thJS} | – | – | 40 | K/W | – |

1) For calculation of R_{thJA} please refer to application note thermal resistance

2.3 ESD Integrity

Table 3 ESD Integrity

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|-----------------------------------|---------------|--------|------|------|------|-----------------------|
| | | Min. | Typ. | Max. | | |
| ESD robustness, HBM ¹⁾ | $V_{ESD-HBM}$ | -1 | – | 1 | kV | All pins |
| ESD robustness, CDM ²⁾ | $V_{ESD-CDM}$ | -500 | – | 500 | V | All pins |

1) According to ANSI/ESDA/JEDEC JS-001 (R = 1.5k Ω , C = 100pF) for Electrostatic Discharge Sensitivity Testing, Human Body Model (HBM)-Component Level

2) According to JEDEC JESD22-C101 Field-Induced Charged Device Model (CDM), Test Method for Electrostatic-Discharge-Withstand Thresholds of Microelectronic Components

2.4 Measured RF Characteristics

2.4.1 Power Supply

Table 4 Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, SPI-Bit 4 = high

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|-----------------------------------|--------------|--------|------|-------|------|-----------------------|
| | | Min. | Typ. | Max. | | |
| Supply voltage | V_{CC} | 3.135 | 3.3 | 3.465 | V | – |
| Supply current | I_{CC} | 110 | 150 | 190 | mA | – |
| Supply voltage temperature sensor | V_{CCTemp} | 3.135 | 3.3 | 3.465 | V | – |
| Supply current temperature sensor | I_{CCTemp} | 1.3 | 2.2 | 3 | mA | – |

2.4.2 TX Section

Table 5 Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high¹⁾

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|---|----------------------------|--------------------|------|-------|-------|--|
| | | Min. | Typ. | Max. | | |
| VCO frequency range | f_{VCO} | 24.0 | – | 24.25 | GHz | – |
| VCO fine tuning voltage ²⁾ | V_F | 0.8 ³⁾ | – | 3.0 | V | 0.8V ≤ V_F ≤ 1.25V for $V_{CMIN}=0.75V$, 1.25V ≤ V_F ≤ 3.0V for $V_{CMIN}=0.75V+1.7V*(V_F-1.25V)/1.75V$, see Figure 2 on Page 12 |
| VCO coarse tuning voltage ²⁾ | V_C | 0.75 ³⁾ | – | 3.0 | V | 0.75V ≤ V_C ≤ 1.35V for $V_{FMIN}=0.8V$, 1.35V ≤ V_C ≤ 3.0V for $V_{FMIN}=V_C-0.55V$, see Figure 2 on Page 12 |
| VCO tuning slope FINE | $\Delta f / \Delta V_F$ | 200 | – | 1000 | MHz/V | 24GHz ≤ f ≤ 24.25GHz see Figure 2 on Page 12 |
| VCO tuning slope COARSE | $\Delta f / \Delta V_C$ | 400 | – | 2000 | MHz/V | 24GHz ≤ f ≤ 24.25GHz see Figure 2 on Page 12 |
| VCO temperature drift | $\Delta f / \Delta T$ | -10 | -6 | 0 | MHz/K | Min @ $T = -40^\circ\text{C}$ |
| VCO pushing | $\Delta f / \Delta V_{CC}$ | -350 | 60 | 350 | MHz/V | Absolute values |

1) Performance based on Application Circuit Figure 3 on Page 16, Cross Section of Application Board, Compensation Structures and Application Board Layout Figure 5 on Page 21ff and Footprint Figure 9 on Page 24

- 2) At tuning pins chipinternal pull-up of $60k\Omega \pm 20\%$ to VCC; max.- and min. temperature tuning voltage limits are chosen in a way that they can be linearly interpolated within operating temperature range
- 3) Specified min. value for temperature $\leq 25^\circ\text{C}$; min. value for temperatures $>25^\circ\text{C}$ is based on following formular $V_{\text{CMIN}} = V_{\text{FMIN}} = 0.8\text{V} + 0.4\text{V} * (T [^\circ\text{C}] - 25^\circ\text{C}) / 100^\circ\text{C}$

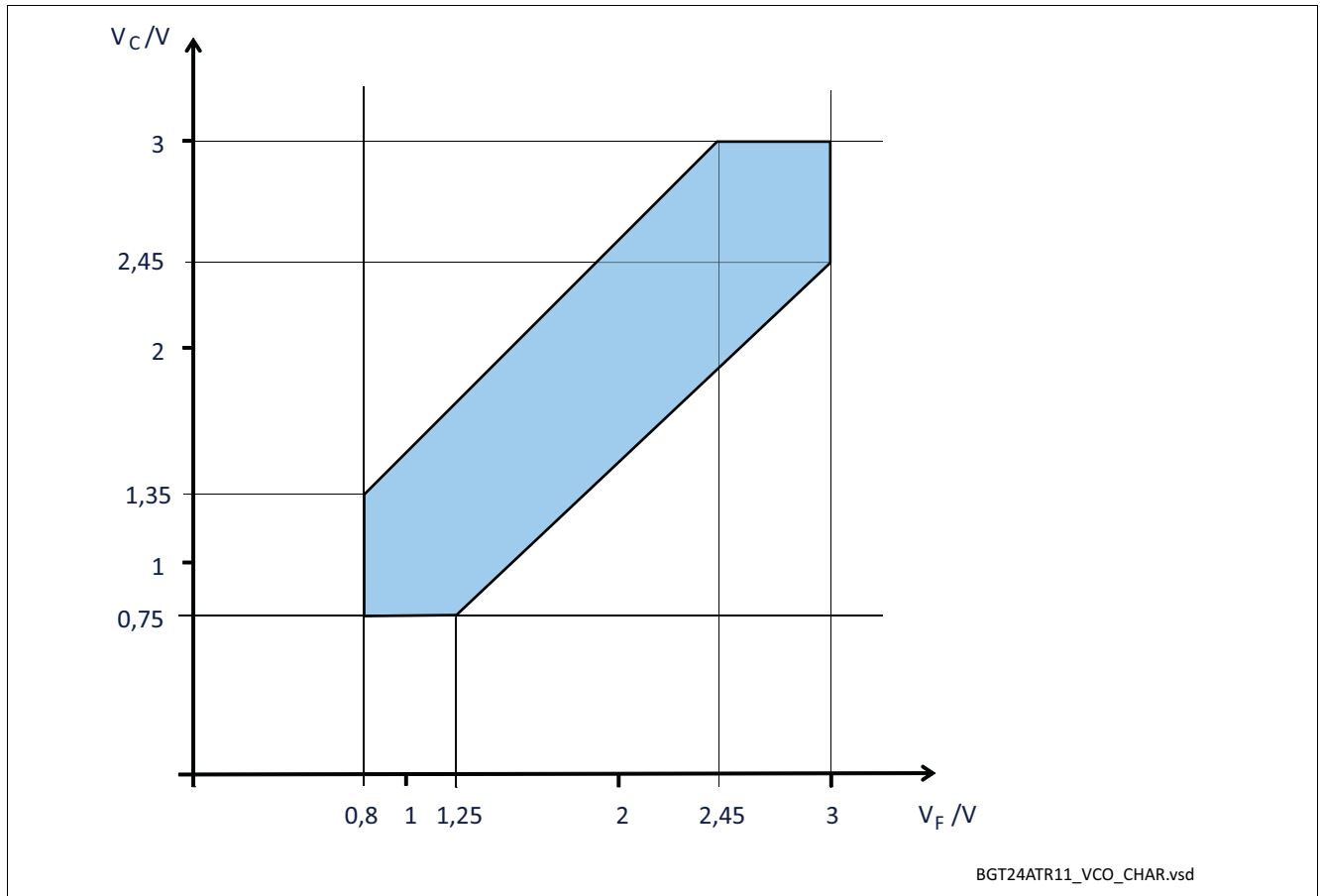


Figure 2 VCO Tuning Window

Table 6 Typical Characteristics $T_A = -40 .. 125^\circ\text{C}$, $f = 24.0 .. 24.25\text{ GHz}$, SPI-Bit 4 = high¹⁾

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|---|-------------------------------------|--------|------------------------|------|----------|--|
| | | Min. | Typ. | Max. | | |
| VCO phase noise | P_N | – | -85 | -75 | dBc/Hz | @ 100kHz offset, $V_F = V_C$ |
| TX/TXX load impedance | Z_{TX} Z_{TXX} | – | 19.8-j20.9 18-j17.3 | – | Ω | Typical value at 24.125GHz and VSWR $\leq 2:1$ |
| Max. TX output power | P_{TX} | 6 | 11 | 15 | dBm | – |
| Max. TX output power for $V_{\text{CC}} = 3.3\text{V} \pm 50\text{mV}$ | P_{TX} | 6.5 | 11 | 14.5 | dBm | – |
| TX output power adjustable range | a_{TX} | 3 | 9 | – | dB | Adjustable via SPI |
| TX output power in “off” mode ²⁾ | P_{TXoff} | – | – | -30 | dBm | Parameter based on IFX eval board design |

Table 6 Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high¹⁾ (cont'd)

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|--|-----------------------------|--------|------------|------|------------|---|
| | | Min. | Typ. | Max. | | |
| LO load impedance | Z_{LO} | – | 24.4-j25.8 | – | Ω | Typical value at 24.125GHz and VSWR \leq 2:1 |
| LO output power ³⁾ | P_{LO} | -8 | 0 | 6 | dBm | SPI-Bit 4 = high |
| Q1 Prescaler division ratio | D_{Q1} | – | 2^4 | – | – | – |
| Q1 Prescaler output power | P_{Q1} | -13 | -9 | -5 | dBm | Q1 loaded with 100 Ohm (AC- coupled) |
| Q1 load impedance | Z_{Q1} | – | 100 | – | Ω | – |
| Q2 Prescaler division ratio | D_{Q2} | – | 2^{20} | – | – | – |
| Q2 Prescaler max. output voltage | $V_{\max Q2}$ | 2.4 | – | – | V | Test condition: Q2 loaded with high impedance probe (1 MOhm, 13 pF) |
| Q2 Prescaler min. output voltage | $V_{\min Q2}$ | – | – | 0.8 | V | Test condition: Q2 loaded with high impedance probe (1 MOhm, 13 pF) |
| Q2 Prescaler max. output source current | $I_{\max \text{source} Q2}$ | 1.2 | – | – | mA | Test condition: Q2 loaded with 50 Ohm to Vcc |
| Q2 Prescaler max. output sink current | $I_{\max \text{sink} Q2}$ | 1.2 | – | – | mA | Test condition: Q2 loaded with 50 Ohm to Vcc |
| Q2 Prescaler output resistance in disable mode | $R_{Q2,DIS}$ | 100 | – | – | k Ω | – |
| Voltage at Txoff for disabling TX output power | $V_{TX,OFF}$ | 1.5 | – | – | V | – |
| Voltage at Txoff for enabling TX output power | $V_{TX,ON}$ | – | – | 0.5 | V | – |
| TXon/off switching time | $t_{ON/OFF}$ | – | – | 500 | ns | – |

- 1) Performance based on Application Circuit Figure 3 on Page 16, Cross Section of Application Board, Compensation Structures and Application Board Layout Figure 5 on Page 21ff and Footprint Figure 9 on Page 24
- 2) Guaranteed by device design
- 3) High LO buffer output power in “high” mode otherwise typ. 4dB reduced LO-output power

2.4.3 RX Section

Table 7 Typical Characteristics $T_A = -40 \dots 125 \text{ }^\circ\text{C}$, $f = 24.0 \dots 24.25 \text{ GHz}$, SPI-Bit 4 = high¹⁾

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|---------------------------------------|---|--------|------------|-------|----------|---|
| | | Min. | Typ. | Max. | | |
| RFIN frequency range | f_{RFIN} | 24.0 | – | 24.25 | GHz | – |
| RFIN port impedance ²⁾ | Z_{RFIN} | – | 22.9-j14.9 | – | Ω | Typical value at 24.125GHz and VSWR \leq 2:1 |
| RFIN VSWR | VSWR | – | – | 2:1 | – | At load port of off-chip compensation network as proposed |
| IF frequency range | f_{IF} | 0 | – | 10 | MHz | – |
| IF output impedance | Z_{IF} | 850 | 1000 | 1150 | Ω | – |
| Leakage LO to RFIN | $L_{\text{LO} \rightarrow \text{RFIN}}$ | – | – | -30 | dBm | Parameter based on IFX eval board design |
| Voltage conversion gain ³⁾ | G_C | 19 | 26 | 31 | dB | $R_{\text{LOAD,IF}} > 10 \text{ k}\Omega$ |
| LNA gain reduction | ΔG_{CLG} | 3 | 5 | 8 | dB | – |
| SSB noise figure | N_{SSB} | – | 12 | 20 | dB | Single sideband at $f_{\text{IF}} = 100 \text{ kHz}$ |
| IF 1/f corner frequency | f_c | – | 10 | 20 | kHz | – |
| Input compression point | $IP_{1\text{dB}}$ | -17 | -12 | – | dBm | – |
| Input 3'rd order intercept point | $IIP3$ | -8 | -4 | – | dBm | – |
| Quadrat. phase imbalance | ε_p | -10 | – | 10 | deg | – |
| Quadrat. amplitude imbalance | ε_A | -1 | – | 1 | dB | – |

1) Performance based on Application Circuit Figure 3 on Page 16, Cross Section of Application Board, Compensation Structures and Application Board Layout Figure 5 on Page 21ff and Footprint Figure 9 on Page 24

2) Guaranteed by device design

3) Lowest gain at high temperature, highest gain at low temperature

2.5 Temperature Sensor

Monitoring of the chip temperature is provided by the on-chip temperature sensor which delivers temperature-proportional voltage to the TEMP output. The temp. sensor can be independently biased through VCCTEMP. Thereby the chip temperature can be monitored while the main supply of the transceiver is switched off.

Table 8 Typical Characteristics Temperature Sensor $T_A = -40 .. 125\text{ }^\circ\text{C}^1$

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|----------------------------|-----------------------|--------|------|----------|------------------|-----------------------|
| | | Min. | Typ. | Max. | | |
| Temperature range | T_{TSENS} | -40 | – | 125 | $^\circ\text{C}$ | – |
| Output temperature voltage | $V_{\text{OUT,TEMP}}$ | – | 1.50 | – | V | @ 25 $^\circ\text{C}$ |
| Sensitivity | S_{TSENS} | – | 4.5 | – | mV/K | – |
| Overall accuracy error | Err_{TSENS} | – | – | ± 15 | K | – |

1) all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

2.6 Power Detector

For RF output power indication, peak voltage detectors are connected to the output of the TX power amplifier and to the LO medium power amplifier. To eliminate temperature and supply voltage variations, a reference output voltage V_{REF} is available through the ANA output for the TX and LO power sensor. The compensated detector output voltage is given by the difference between V_{OUT} and V_{REF} for both power sensors respectively. This voltage is proportional to the RF voltage swing at the individual amplifier outputs, its characteristic is non-directional.

Table 9 Typical Characteristics Power Detector $T_A = -40 .. 125\text{ }^\circ\text{C}$, $V_{\text{CC}} = 3.3\text{ V}^1$

| Parameter | Symbol | Values | | | Unit | Note / Test Condition |
|------------------------|---|--------|------|------|------|-----------------------------------|
| | | Min. | Typ. | Max. | | |
| Power range | P_{PSENS} | -10 | – | 15 | dBm | – |
| TX power sensor output | $V_{\text{OUT,TX}} - V_{\text{REF,TX}}$ | – | 550 | – | mV | @ $P_{\text{TX}} = 11\text{ dBm}$ |
| LO power sensor output | $V_{\text{OUT,LO}} - V_{\text{REF,LO}}$ | – | 50 | – | mV | @ $P_{\text{LO}} = 0\text{ dBm}$ |

1) all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

3 Application Circuit and Block Diagram

3.1 Application Circuit Schematic

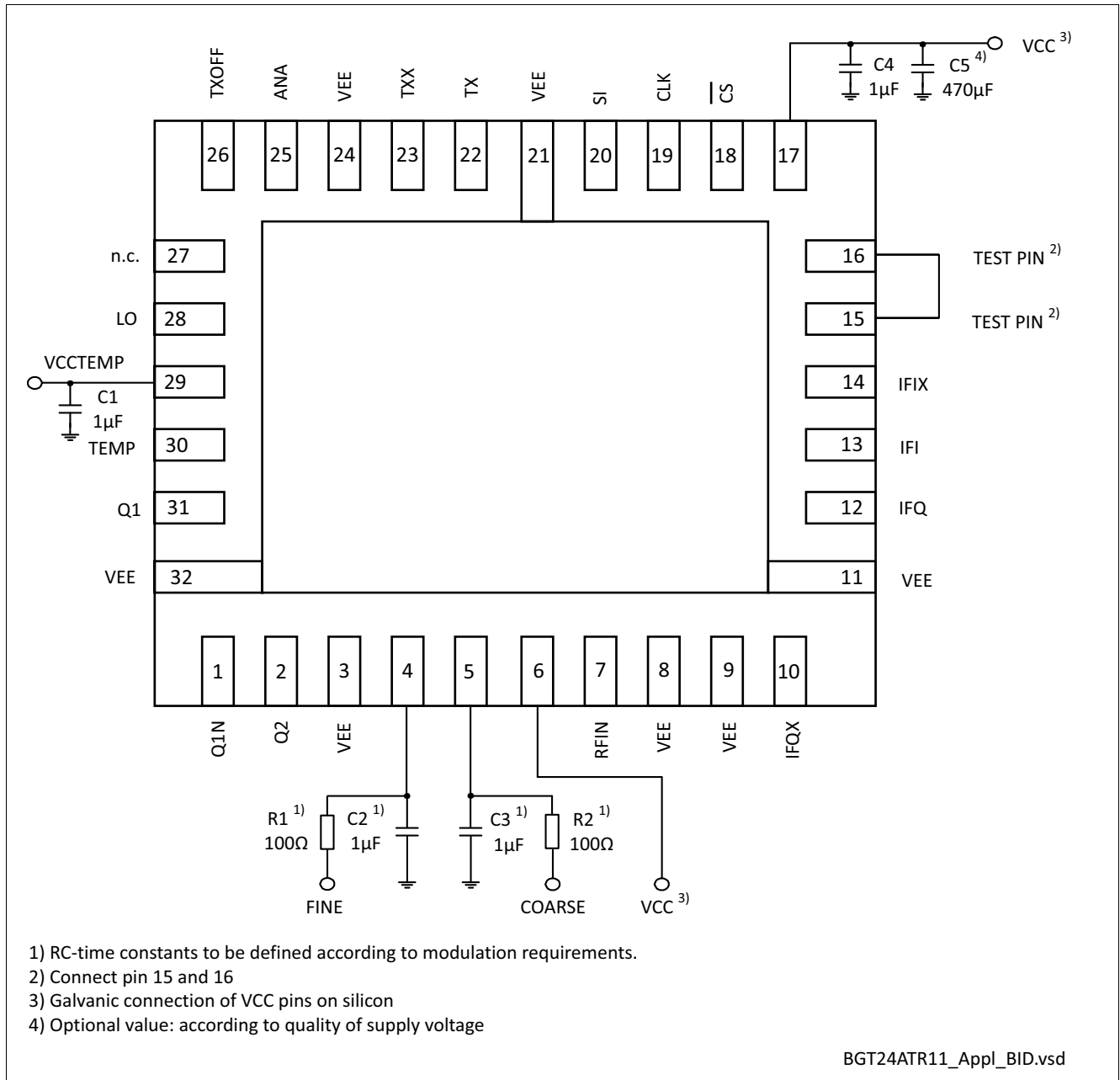


Figure 3 Application Circuit with Chip Outline (Top View)

Table 10 Bill of Materials

| Part Number | Part Type | Manufacturer | Size | Comment |
|-------------|----------------|--------------|---------|---------|
| C1 ... C5 | Chip capacitor | Various | Various | – |
| R1 ... R2 | Chip resistor | Various | 0402 | – |

3.2 Pin Description

Table 11 Pin Definition and Function

| Pin No. | Name | Function |
|---------|------------------------|---|
| 1 | Q1N | Complementary prescaler output 1.5GHz |
| 2 | Q2 | Prescaler output 23kHz |
| 3 | VEE | Ground |
| 4 | FINE | VCO fine tuning input |
| 5 | COARSE | VCO coarse tuning input |
| 6 | VCC | Supply voltage; Total current divided equal on both VCC pins |
| 7 | RFIN | RF input downconverter |
| 8 | VEE | Ground |
| 9 | VEE | Ground |
| 10 | IFQX | Complementary quadrature phase IF output downconverter |
| 11 | VEE | Ground |
| 12 | IFQ | Quadrature phase IF output downconverter |
| 13 | IFI | In phase IF output downconverter |
| 14 | IFIX | Complementary in phase IF output downconverter |
| 15 | TEST PIN | Test pin; DC coupled pin |
| 16 | TEST PIN | Test pin; DC coupled pin |
| 17 | VCC | Supply voltage; Total current divided equal on both VCC pins |
| 18 | $\overline{\text{CS}}$ | Chip select input SPI (inverted) |
| 19 | CLK | Clock input SPI interface |
| 20 | SI | Data input SPI interface |
| 21 | VEE | Ground |
| 22 | TX | Transmit output |
| 23 | TXX | Complementary transmit output |
| 24 | VEE | Ground |
| 25 | ANA | Analog output |
| 26 | TXOFF | Pulsable Pin / Please connect to VEE in case TXOFF function is controlled via SPI |
| 27 | n.c. | Not connected |
| 28 | LO | LO output |
| 29 | VCCTEMP | Temperature sensor supply voltage |
| 30 | TEMP | Temperature sensor output |
| 31 | Q1 | Prescaler output 1.5GHz |
| 32 | VEE | Ground |

3.3 SPI

1.) Three signals control the serial peripheral interface of the BGT24ATR11:

SI (Data); CLK (Clock); \overline{CS} (Chip select)

2.) The data bits SI (MSB first) are read in the shift with falling edge of the CLK signal.

Please make sure, that the data is present at least 10 ns before and at least 10 ns after the falling edge of the clock signal.

3.) The CLK and \overline{CS} signals are combined internally.

At least 20 ns before first rising edge of the first CLK signal \overline{CS} needs to be in "low" state.

While the Data is read, \overline{CS} has to remain in "low" state.

4.) When Data read in is finished, the shift register content will be written in the latch at the rising edge of the \overline{CS} signal. The time between the last falling edge of the CLK signal and the rising edge of the \overline{CS} must be at least 20 ns.

Table 12 SPI Data Bit Description

| Data Bit | Name | Description (Logic High) | Power ON State |
|----------|------------|---|----------------|
| 15 (MSB) | GS | LNA Gain reduction | low |
| 14 ..13 | – | Not used | low |
| 12 | DIS_PA | TX power disabled, in case TXon/off function is controlled via TXOFF pin, this bit needs to be set in low state | high |
| 11 | AMUX2 | Analog multiplexer control bit 2 | high |
| 10 | Test Bit | Test bit, must be low otherwise malfunction | low |
| 9 | Test Bit | Test bit, must be low otherwise malfunction | low |
| 8 | AMUX1 | Analog multiplexer control bit 1 | low |
| 7 | AMUX0 | Analog multiplexer control bit 0 | low |
| 6 | DIS_DIV64k | Disable 64k divider | low |
| 5 | DIS_DIV16 | Disable 16 divider | low |
| 4 | PC2_BUF | High LO buffer output power in "high" mode otherwise typ. 4dB reduced LO-output power | low |
| 3 | PC1_BUF | High TX buffer output power | low |
| 2 | PC2_PA | TX power reduction bit 2 | high |
| 1 | PC1_PA | TX power reduction bit 1 | high |
| 0 | PC0_PA | TX power reduction bit 0 | high |

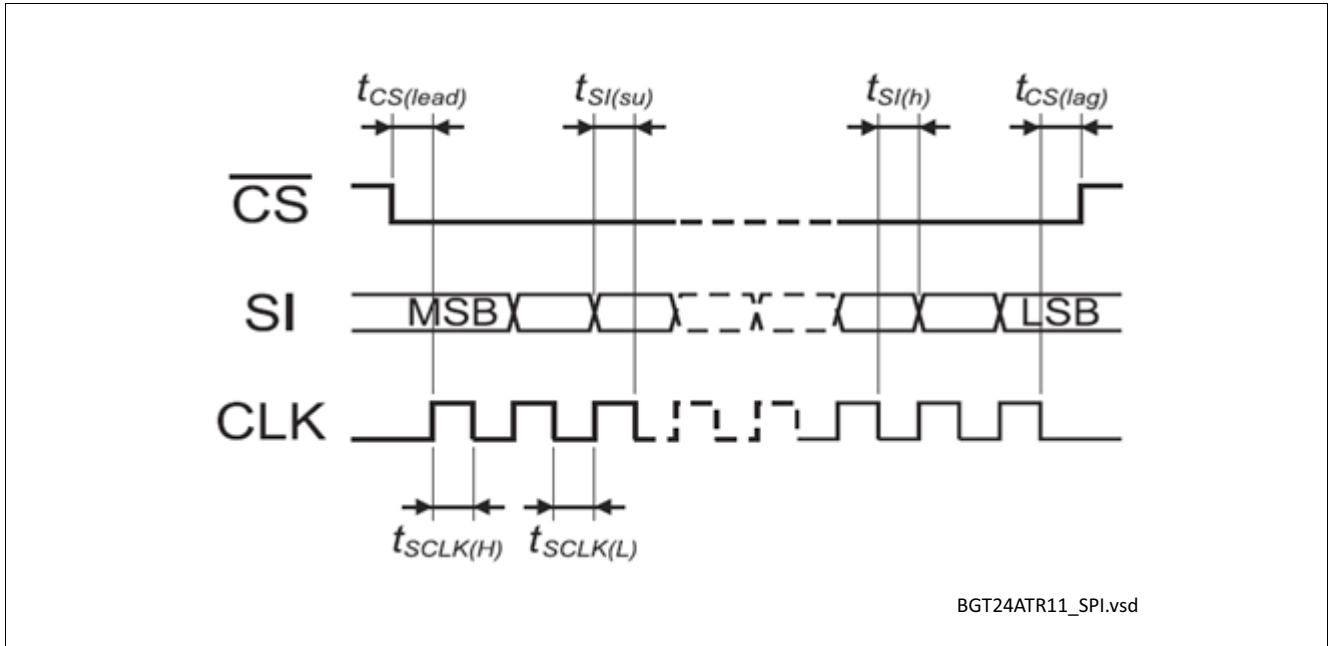


Figure 4 Timing Diagram of the SPI

Table 13 SPI Timing and Logic Levels

| Parameter | Symbol | Values | | | Unit |
|---|----------------|--------|------|----------|---------|
| | | Min. | Typ. | Max. | |
| Serial clock frequency | f_{SCLK} | 0 | – | 50 | MHz |
| Serial clock high time | $f_{SCLK(H)}$ | 10 | – | – | ns |
| Serial clock low time | $t_{SCLK(L)}$ | 10 | – | – | ns |
| Chip select lead time | $t_{CS(lead)}$ | 20 | – | – | ns |
| Chip select lag time | $t_{CS(lag)}$ | 20 | – | – | ns |
| Data setup time | $t_{SI(su)}$ | 10 | – | – | ns |
| Data hold time | $t_{SI(h)}$ | 10 | – | – | ns |
| Low level (SI, CLK, \overline{CS}) | $V_{IN(L)}$ | 0 | – | 0.8 | V |
| High level (SI, CLK, \overline{CS}) | $V_{IN(H)}$ | 2.0 | – | V_{CC} | V |
| Input capacitance (SI, CLK, \overline{CS}) | C_{IN} | – | – | 2 | pF |
| Input current (SI, CLK, \overline{CS}) | I_{IN} | -150 | – | 150 | μA |

Table 14 Truth Table AMUX

| Output signal ANA | AMUX2 | AMUX1 | AMUX0 |
|-------------------|-------|-------|-------|
| $V_{OUT,TX}$ | low | low | low |
| $V_{REF,TX}$ | low | low | high |
| $V_{OUT,LO}$ | low | high | low |
| $V_{REF,LO}$ | low | high | high |
| V_{TEMP} | high | low | low |
| Test_Signal1 | high | low | high |

Table 14 Truth Table AMUX (cont'd)

| Output signal ANA | AMUX2 | AMUX1 | AMUX0 |
|--------------------------|--------------|--------------|--------------|
| Test_Signal2 | high | high | low |
| Test_Signal2 | high | high | high |

3.4 Application Board and Reflow Profile

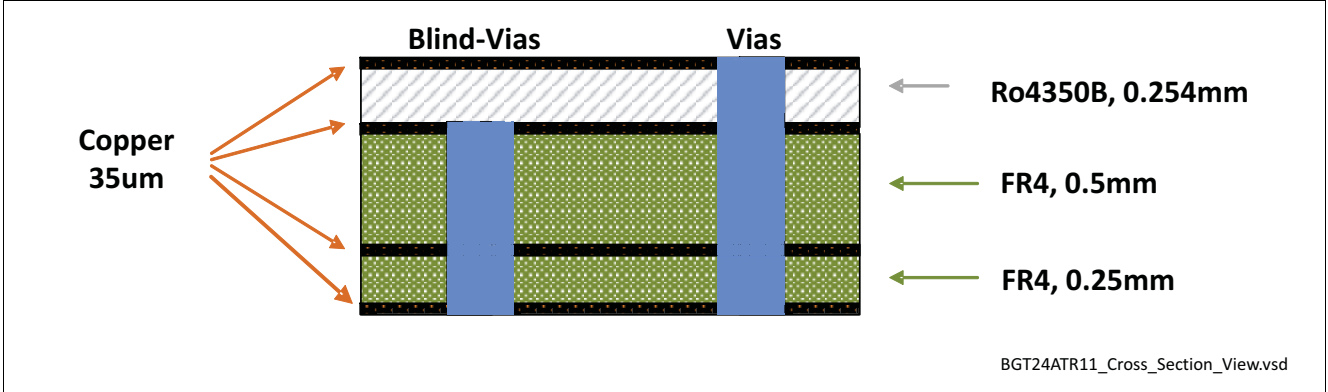


Figure 5 Cross-Section View of Application Board

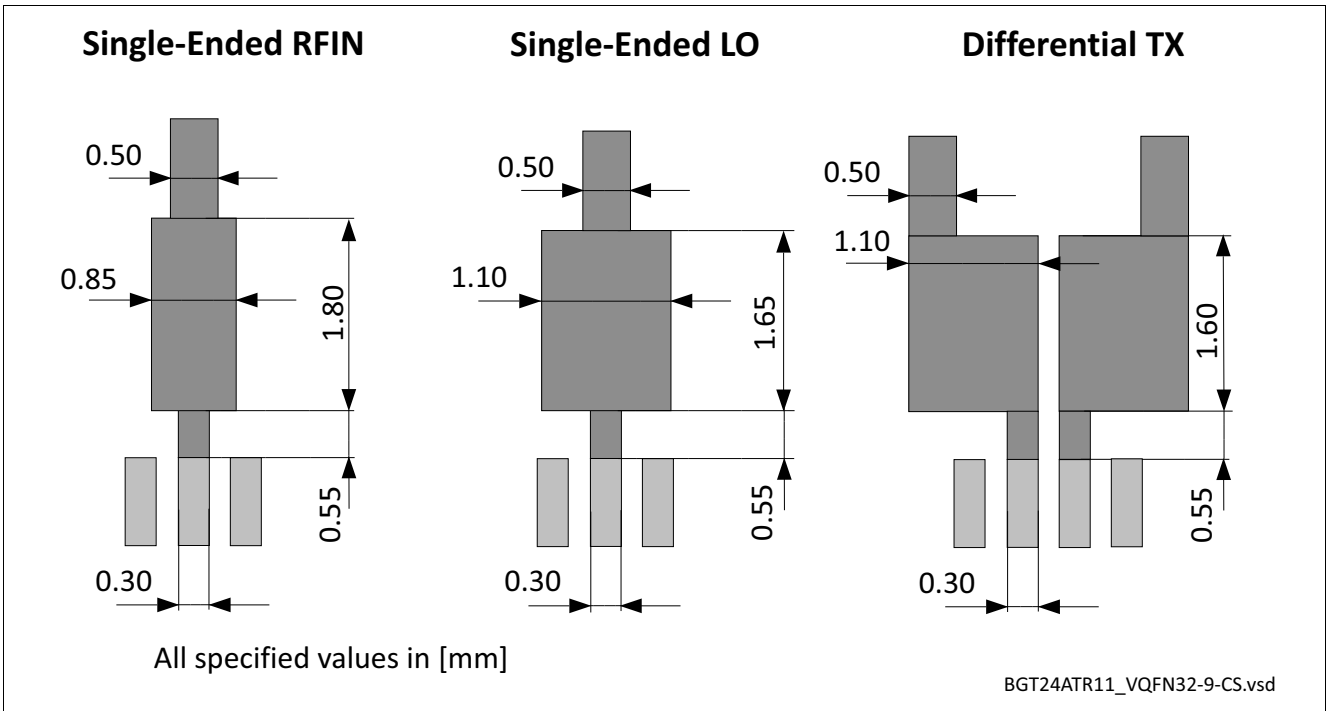


Figure 6 Detail of Compensation Structure (valid for appl. board mat. Ro4350B, 0.254mm acc. to Fig. 5)

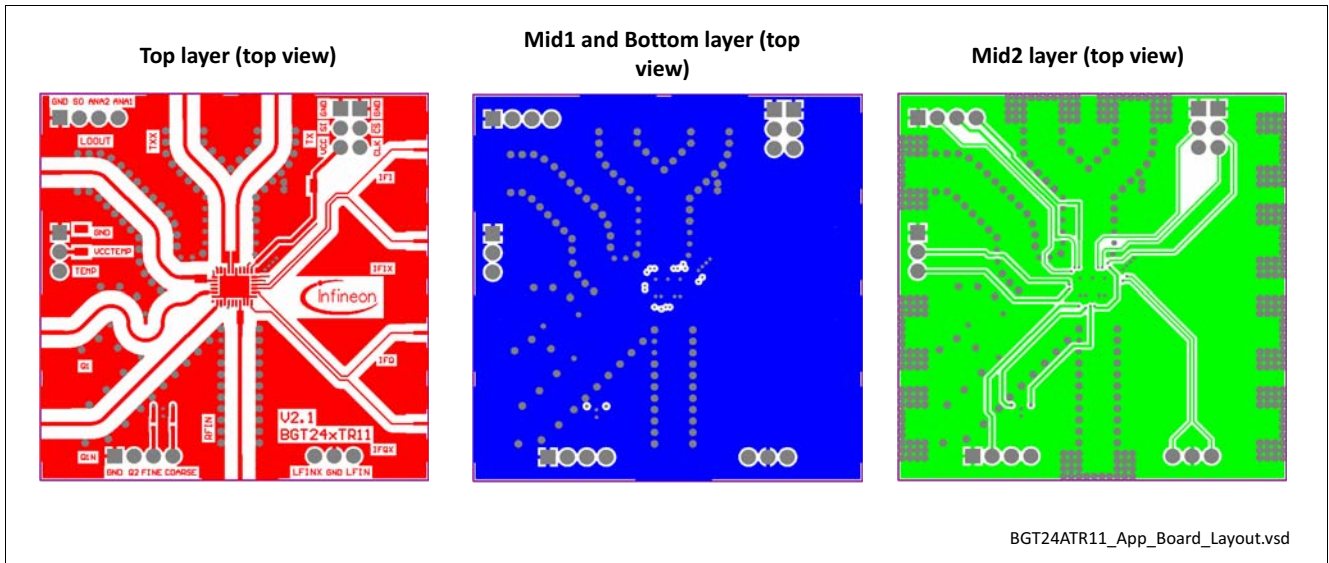


Figure 7 Application Board Layout

Note: In order to achieve the same performance as given in this datasheet please follow the suggested PCB-layout. The compensation structure is critical for RF performance. Via holes as recommended on one of next pages (not shown above).

3.5 Equivalent Circuit Diagram of MMIC Interfaces

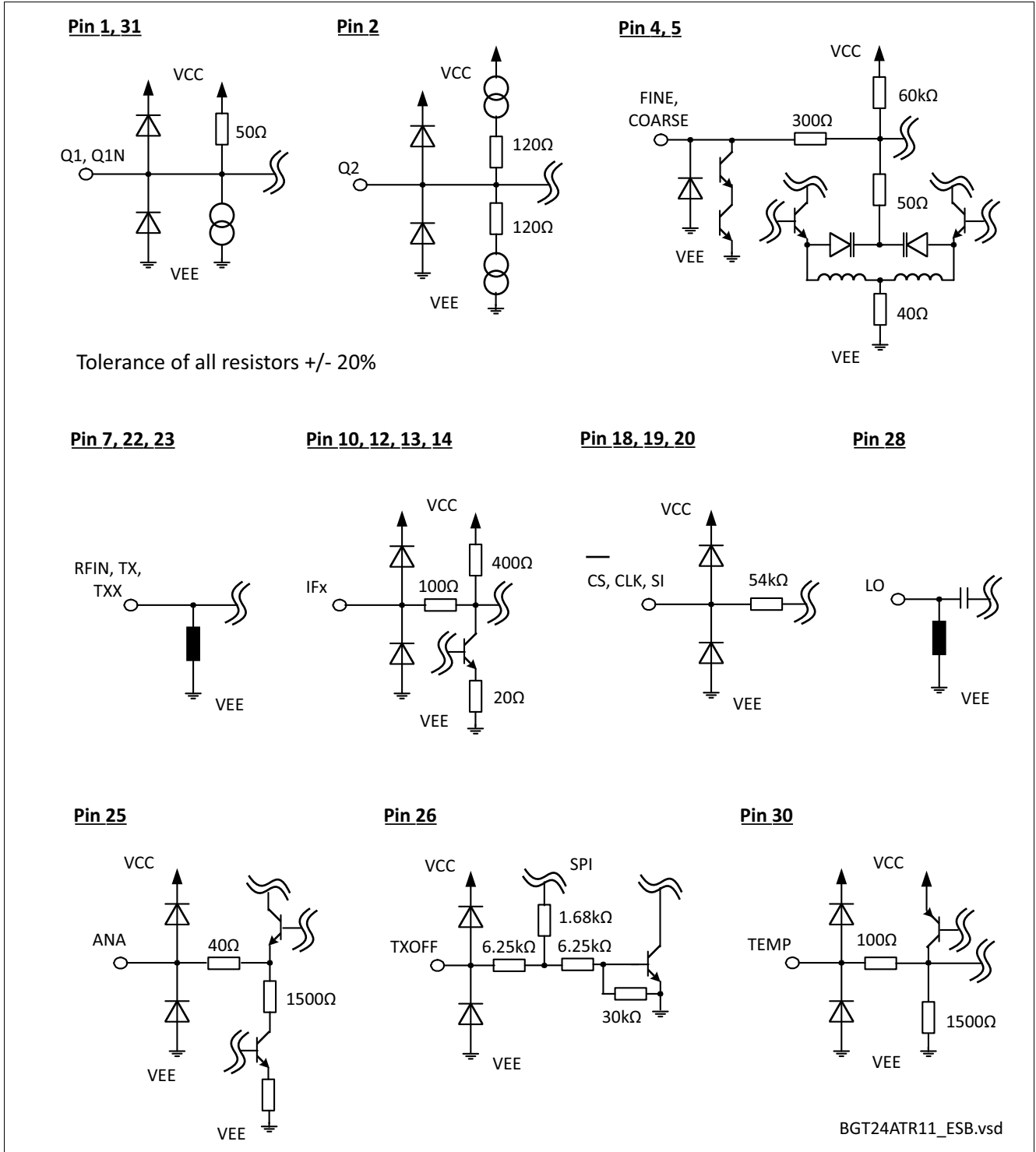


Figure 8 Equivalent Circuit Diagram of MMIC Interfaces

4 Physical Characteristics

4.1 Package Footprint

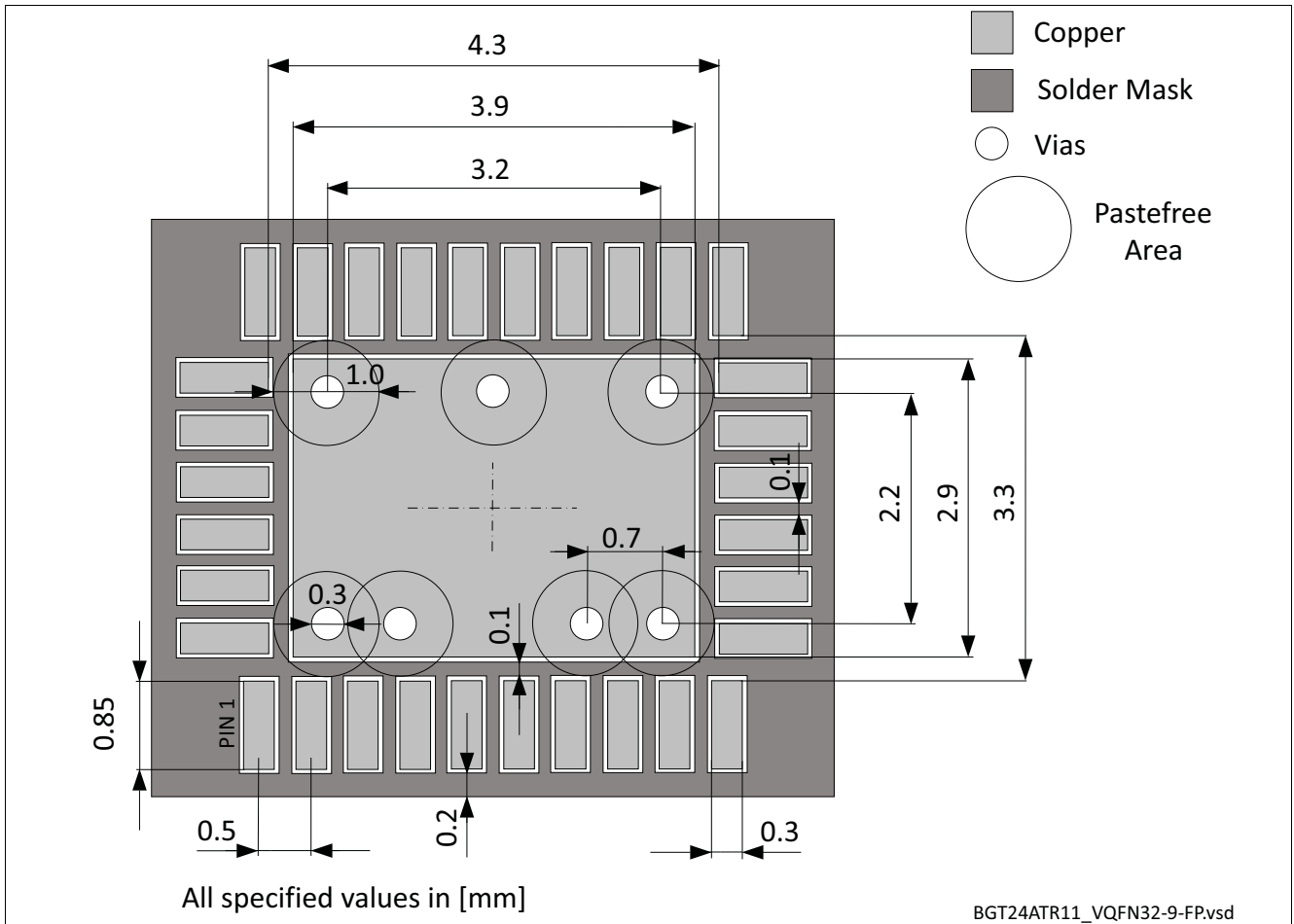


Figure 9 Recommended Footprint and Stencil Layout for the VQFN32-9 Package

4.2 Reflow Profile

Soldering process qualified during qualification with "Preconditioning MSL-3: 30°C. 60%r.h., 192h, according to JEDEC JSTD20".

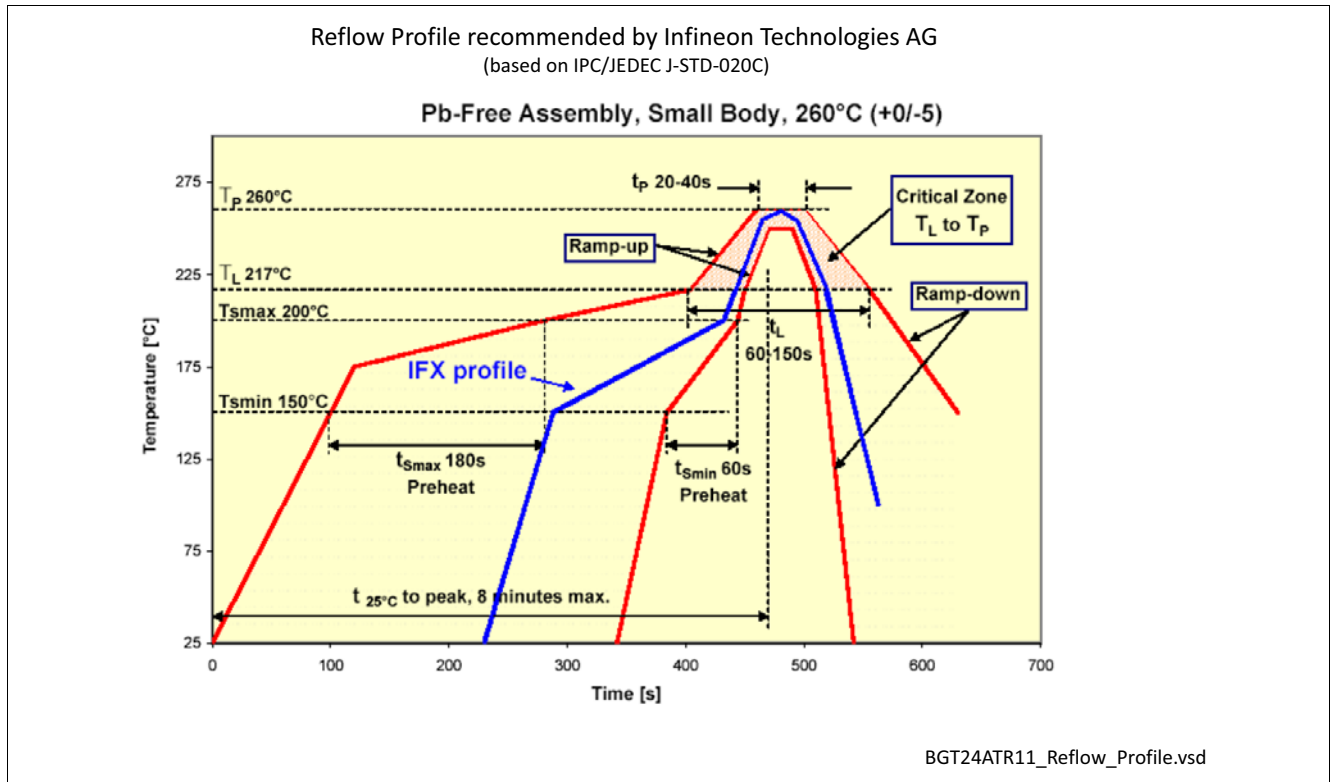


Figure 10 Reflow Profile for BGT24ATR11 (VQFN32-9)

4.3 Package Dimensions

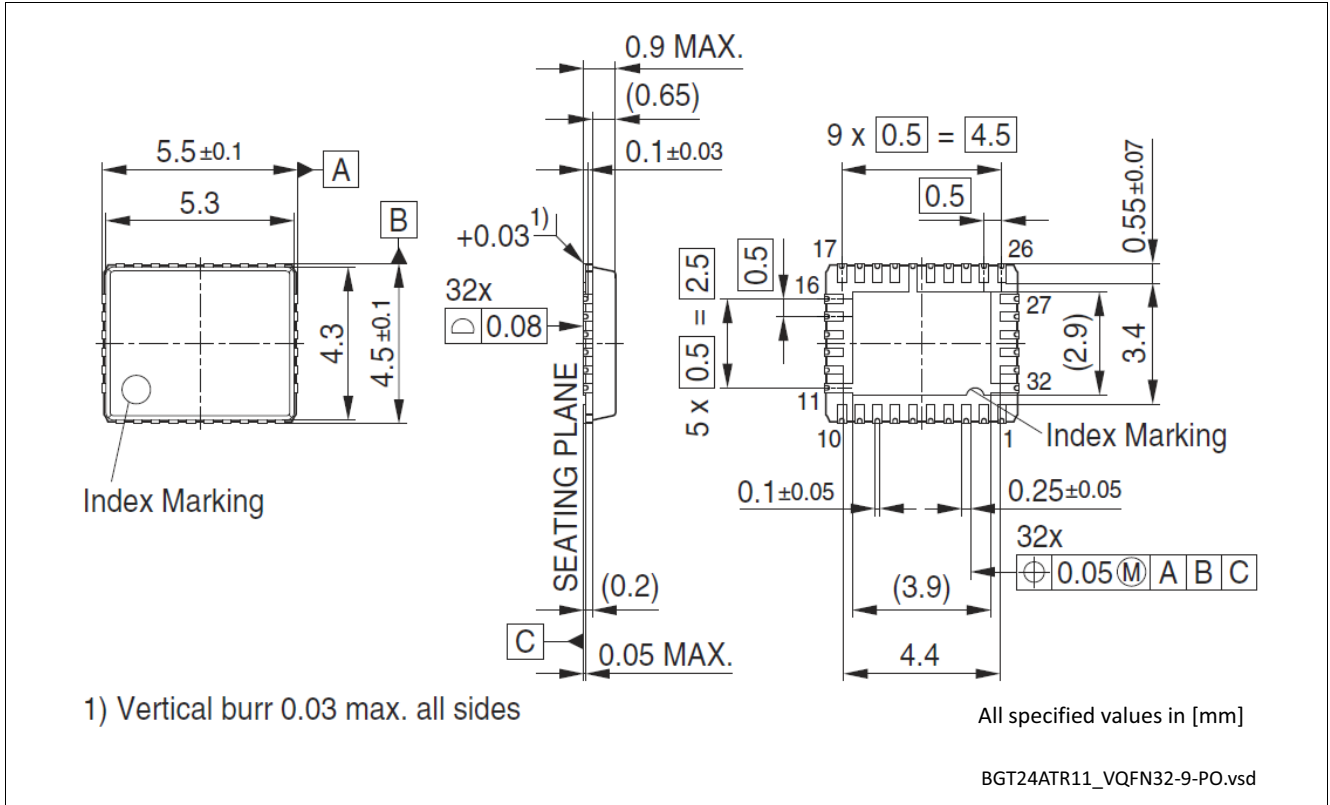


Figure 11 Package Outline (Top, Side and Bottom View) of VQFN32-9

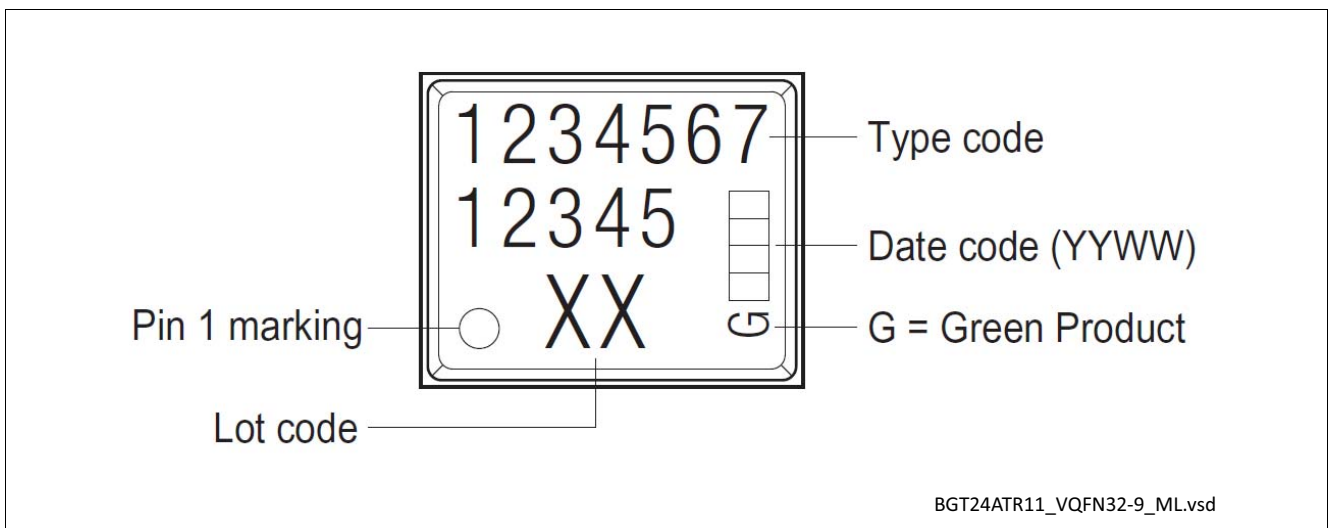


Figure 12 Marking Layout VQFN32-9

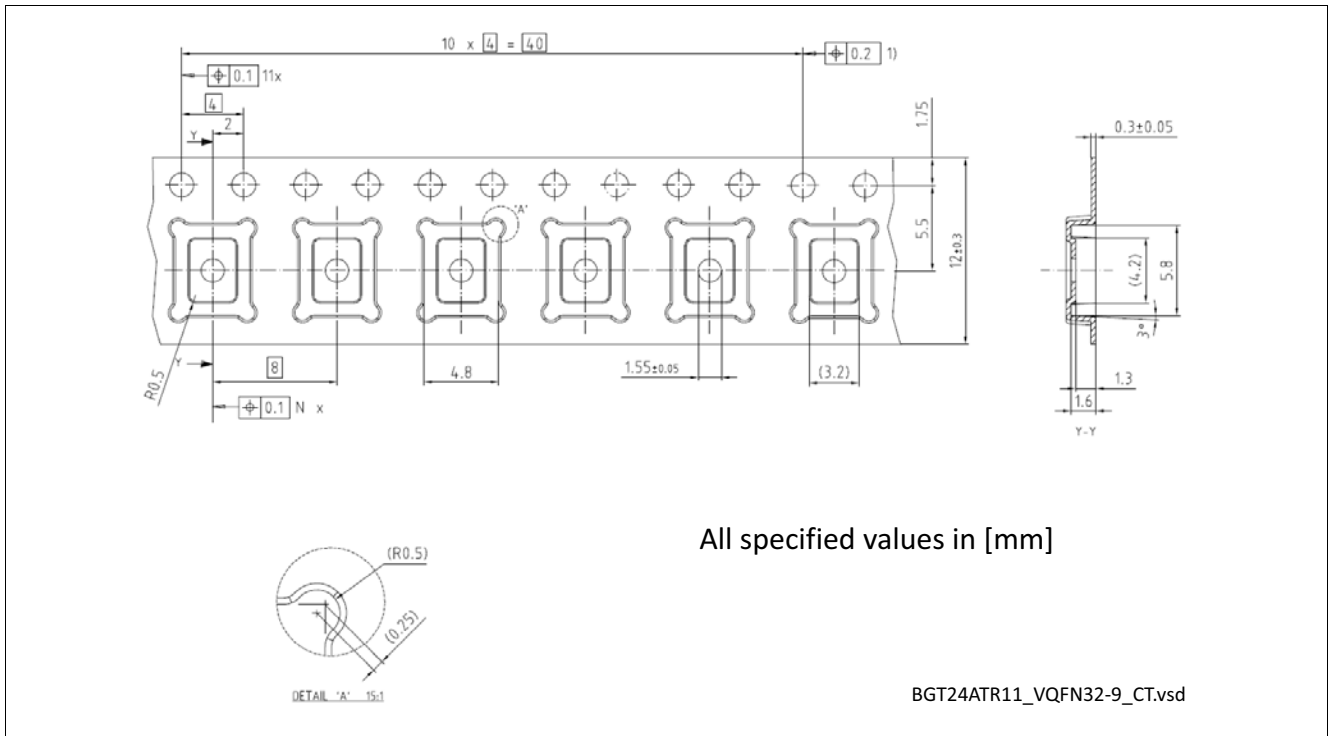


Figure 13 Tape of VQFN32-9

www.infineon.com

Published by Infineon Technologies AG